

1) The current-fed converter shown in Fig. 1 is operating at $f_{sw}=1$ MHz (assume ideal MOSFETs and diodes).

a) Derive the DC transfer function V_O/I_{IN} .

Consider the following specifications: $I_{IN} = 5$ A, $V_O = 5$ V, $P_O = 10$ W.

b) Select the output capacitor C such that the peak-to-peak voltage ripple is less than 1% of the nominal voltage.

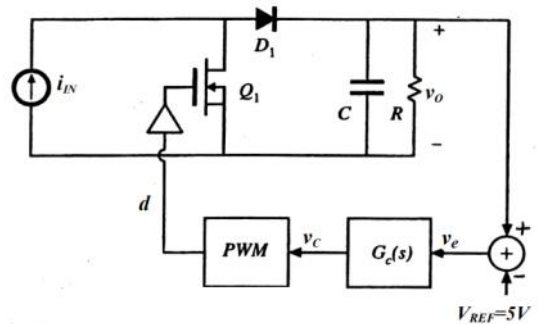
a) ON

$$-I_D D + (I_{IN} - I_D)(1-D) = 0$$

$$-I_D D + I_N(1-D) - I_D + I_D D = 0$$

$$I_D = \frac{V_O}{R} \quad I_N(1-D) - \frac{V_O}{R} = 0$$

$$\frac{V_O}{I_N} = R(1-D)$$

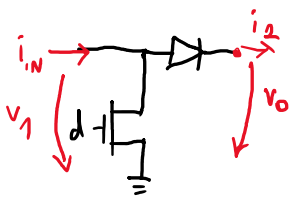


b) $i = C \frac{\Delta V}{\Delta t}$

$$\frac{i \Delta t}{C} \leq 0.01 V_O \Rightarrow C \geq \frac{I_D D T_s}{0.01 \cdot V_O} \Leftrightarrow C \geq \frac{2 \cdot 0.5 \cdot 1/1M}{0.01 \cdot 5} \Leftrightarrow C \geq 24 \mu F$$

c) Derive the transfer functions $G_{oc}(s) = \frac{\tilde{V}_O}{\tilde{V}_C}$ and $G_{vi}(s) = \frac{\tilde{V}_O}{\tilde{I}_{IN}}$, assuming $G_{PWM}(s) = 0.5 \text{ V}^{-1}$ (hint:

perturb and linearize the average diode current $\langle i_d \rangle_{T_s}$).



$$i_2 = (1-d) i_{IN}$$

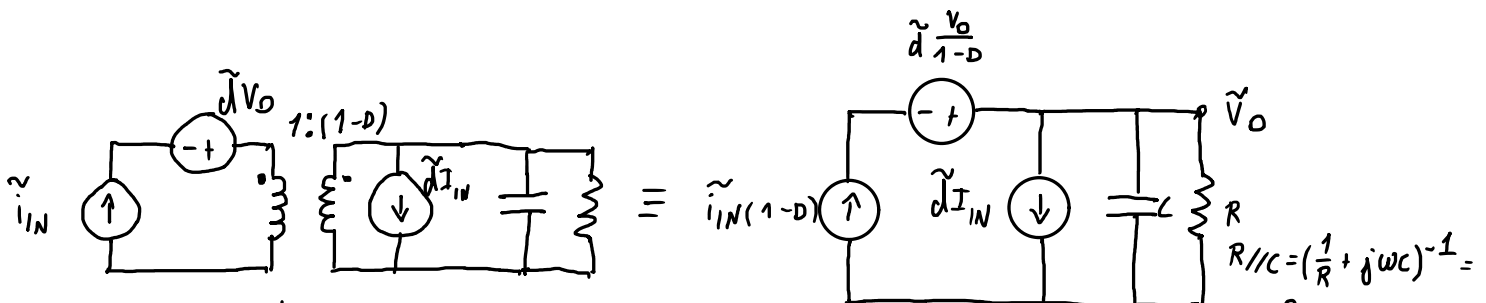
$$V_1 = (1-d) V_O$$

$$I_2 + \tilde{i}_2 = (1-D-\tilde{d})(I_{IN} + \tilde{i}_{IN})$$

$$V_1 + \tilde{v}_1 = (1-D-\tilde{d})(V_O + \tilde{v}_O)$$

$$I_2 = I_{IN}(1-D) \quad \tilde{i}_2 = (1-D)\tilde{i}_{IN} - \tilde{d} I_{IN}$$

$$V_1 = V_O(1-D) \quad \tilde{v}_1 = (1-D)\tilde{v}_O - \tilde{d} V_O$$

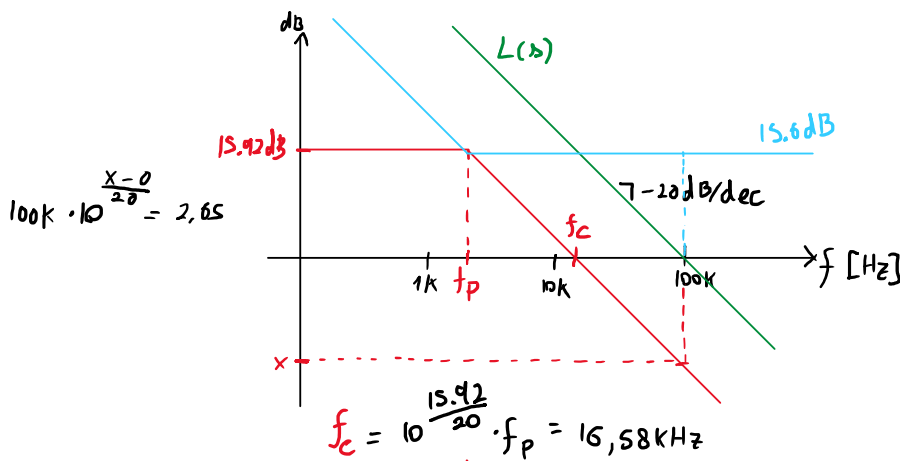


$$G_{oc}(s) = \frac{\tilde{V}_O}{\tilde{V}_C} = \frac{\tilde{V}_O}{\tilde{d}} \cdot \frac{\tilde{d}}{\tilde{V}_C} \Rightarrow \frac{\tilde{V}_O}{\tilde{d}} \bigg|_{\tilde{i}_{IN}=0} = -\frac{R \cdot I_{IN}}{1+sRC} \Rightarrow G_{oc}(s) = -\frac{1}{2} \cdot \frac{R \cdot I_{IN}}{1+sRC} = \frac{R}{1+sRC}$$

$$G_{PWM}(s) = \frac{\tilde{d}}{\tilde{V}_C} = 0.5 \text{ V}^{-1}$$

$$G_{vi}(s) = \frac{\tilde{V}_O}{\tilde{i}_{IN}} \bigg|_{\tilde{d}=0} = \frac{(1-D)R}{1+sRC}$$

d) Determine the compensator transfer function $G_c(s)$ such that the loop gain has a constant slope of -20 dB / dec , crossing the 0 dB axis at $f_c = 100 \text{ kHz}$.



$$G_{oc}(s) = -\frac{1}{2} \cdot \frac{R \cdot I_{IN}}{1 + sRC}$$

$$|G_{oc}(s)| = \frac{1}{2} R I_{IN} = 15.92 \text{ dB}$$

$$f_p = \frac{1}{2\pi RC} = 2,65 \text{ kHz} = f_z$$

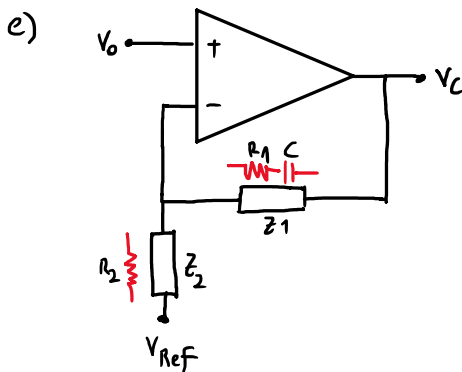
$$L(s) = G_{oc}(s) \cdot G_s(s)$$

$$G_s(s) = K_c \cdot \frac{1 + s/\omega_z}{s} \quad K_c \approx 100k$$

$$L(s) = K_c \cdot \frac{1 + s/\omega_z}{s} \cdot -\frac{1}{2} \cdot \frac{R \cdot I_{IN}}{1 + sRC} = -\frac{1}{2} \cdot \frac{K_c \cdot R \cdot I_{IN}}{s} = -\frac{6.28 \times 10^5}{s}$$

e) Design the error amplifier.

f) Sketch the asymptotic plot of the closed-loop output impedance.



$$\frac{\tilde{V}_c}{\tilde{V}_0} = 1 + \frac{Z_1}{Z_2} = \frac{Z_2 + Z_1}{Z_2} \leadsto K_c \cdot \frac{1 + s/\omega_z}{s}$$

$$Z_1 = R_1 + \frac{1}{sC} \Rightarrow \frac{R_2 + R_1 + \frac{1}{sC}}{R_2} = \frac{sC(R_2 + R_1) + 1}{sC R_2}$$

$$\frac{1}{K_c} = C R_2$$

$$C = 10 \text{ mF} \quad R_2 = 1 \text{ k}\Omega \quad R_1 = 5 \text{ k}\Omega$$

$$\frac{1}{\omega_z} = C(R_2 + R_1)$$

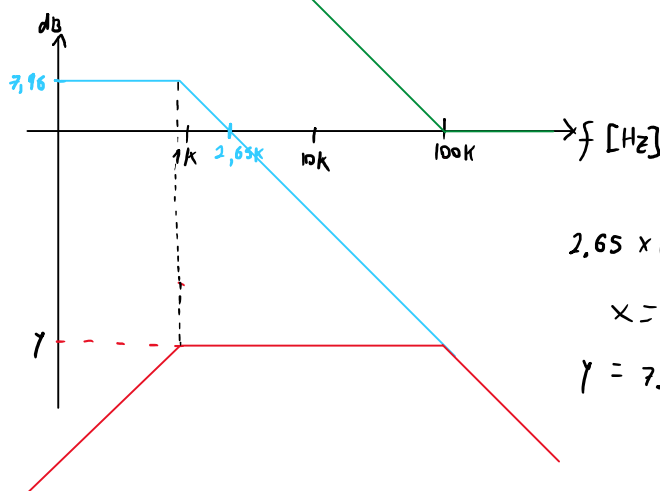
$$f) \quad Z_o^{OL} = \frac{R}{1 + sRC} \quad C = 24 \text{ nF} \quad R = 2.5 \text{ }\Omega$$

$$Z_o^{OL}(0) = R = 7.96 \text{ dB}$$

$$f_{OL} = \frac{1}{2\pi RC} = 2,65 \text{ kHz}$$

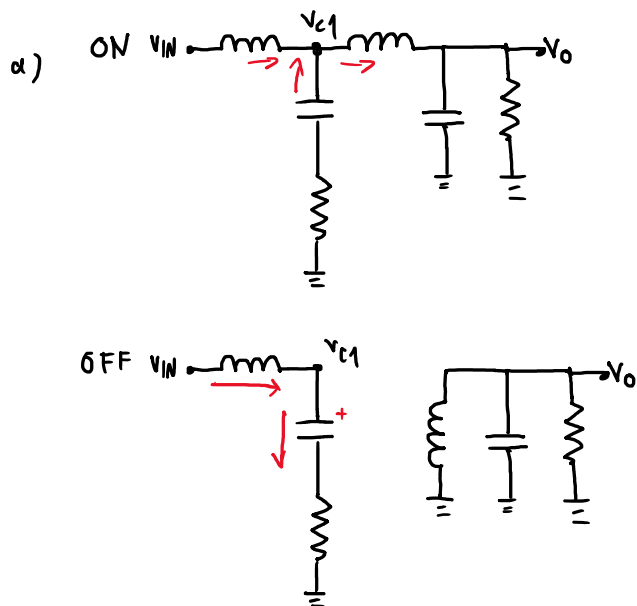
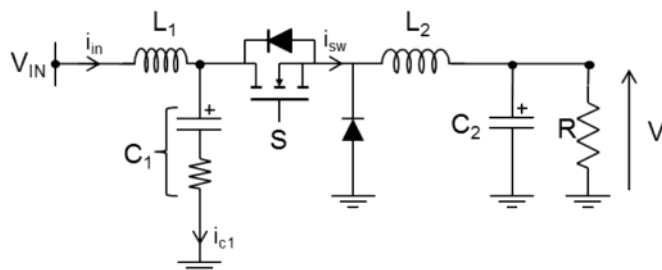
$$1 + L(s) = 6.28 \times 10^5 \cdot \frac{s}{s^2 + 6.28 \times 10^5 s + 1}$$

$$Z_o^{CL} = \frac{Z_o^{OL}}{1 + L(s)}$$



2) The buck converter shown in Fig. 1 operates in CCM at $f_{SW} = 500\text{kHz}$. The current ripple in L_2 is negligible. Assuming $V_{IN} = 12\text{V}$, $V_O = 5\text{V}$, $P_O = 15\text{W}$, $C_1 = 270\mu\text{F}$ ($\text{ESR} = 22\text{m}\Omega$, $V_{\text{rated}} = 25\text{V}$, $I_{\text{rated}} = 1520\text{mA}_{\text{rms}}$):

- draw a plot of i_{C1} as a function of time;
- calculate the voltage ripple ΔV_{C1} ;
- calculate the average power dissipated by C_1 ;
- select the inductance L_1 such that the peak-to-peak current ripple Δi_{L1} is less than 6mA .



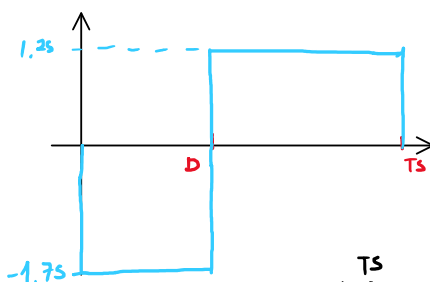
$$(V_{IN} - V_{C1})D + (V_{IN} - V_{C1})(1-D) = 0 \Rightarrow V_{IN} = V_{C1}$$

$$(V_{C1} - V_O)D - V_O(1-D) = 0 \Leftrightarrow V_{IN}D - V_O = 0 \Rightarrow$$

$$\frac{V_O}{V_{IN}} = D = \frac{5}{12} = 0.4166$$

$$\text{ON } I_{L1} + I_L = I_O \Rightarrow I_{L1} = 1.75\text{A}$$

$$I_{L1} = I_{IN} = DI_O = 1.25\text{A} \quad \text{OFF}$$



$$b) \Delta V_{C1} = \Delta V_C + \Delta V_{\text{ESR}} = 71.4\text{mV}$$

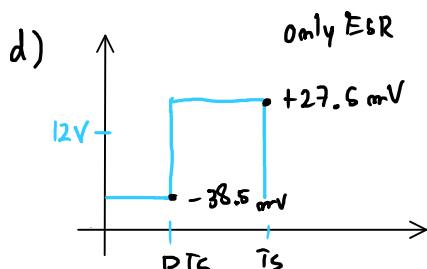
$$\Delta V_{\text{ESR}} = \text{ESR} \cdot \Delta I = 22 \cdot 10^{-3} \cdot (1.25 + 1.75) = 66\text{mV}$$

$$\Delta V_C = \frac{I_O D (1-D) T_S}{C} = 5.4\text{mV}$$

$$c) I_{C1}^2 \text{rms} = \frac{1}{T_S} \int_0^{T_S} I_{C1}^2 dt = \frac{1}{T_S} \cdot \left[\int_0^{DT_S} (I_O(1-D))^2 dt + \int_{DT_S}^{T_S} (I_O \cdot D)^2 dt \right] =$$

$$= I_O^2 (1-D)^2 D + I_O^2 D^2 (1-D) = I_O^2 D (1-D)$$

$$P_{C1} = (I_{C1} \text{rms})^2 \cdot \text{ESR} = I_O^2 D (1-D) \text{ESR} = 48.125\text{mW}$$



$$V_{L1} = V_{IN} - V_{C1}$$

$$V_{L1}|_{\text{ON}} = -38.5\text{mV} - \left(\frac{\Delta V_C}{2}\right) \approx -41.2\text{mV}$$

$$V_{L1}|_{\text{OFF}} = 27.5\text{mV} + \left(\frac{\Delta V_C}{2}\right) \approx 30.2\text{mV}$$

$$\Delta i_{L1} = \frac{V_{L1}|_{\text{ON}} \cdot DT_S}{L_1}$$

$$L_1 \geq \frac{V_{L1}|_{\text{ON}} DT_S}{\Delta i_{L1}} = 5.347\mu\text{H}$$

$$L_1 \geq 5.722\mu\text{H}$$